

# Phase Noise and its Measurements

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# Agenda

- **Phase Noise Basics**
  - What is Phase Noise  $\mathcal{L}(f)$  ?
  - Spectral Density of Phase Fluctuations  $S_{\phi}(f)$ 
    - Relations between SSB PN and RMS phase deviation
  - Jitter and Phase Noise
    - Relations between Jitter and RMS phase deviation
- Phase Noise Measurements
  - Direct-Spectrum Method
  - Phase Detector Method
  - Cross-correlation of 2-Channel Measurements
- Keysight Phase Noise Measurement Solutions



# What is Phase Noise?

Ideal versus real-world signals

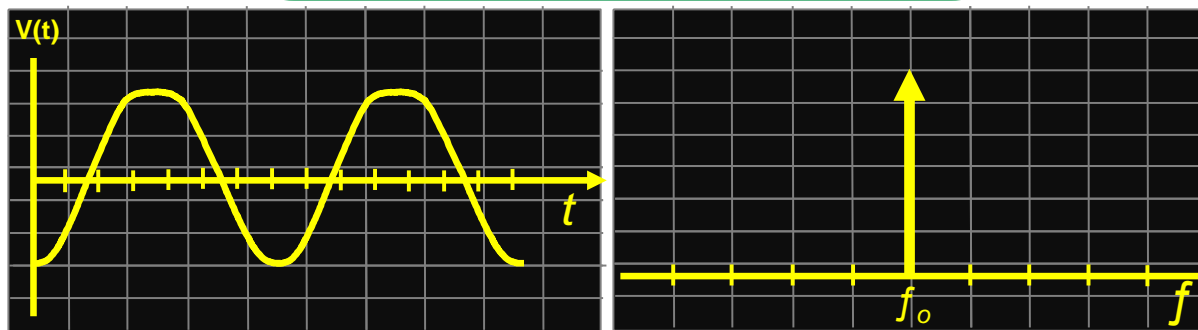
## Ideal sinusoidal signal

$$V(t) = A_o \sin(2\pi f_o t)$$

where

$A_o$  = nominal amplitude

$f_o$  = nominal frequency



Time

Frequency

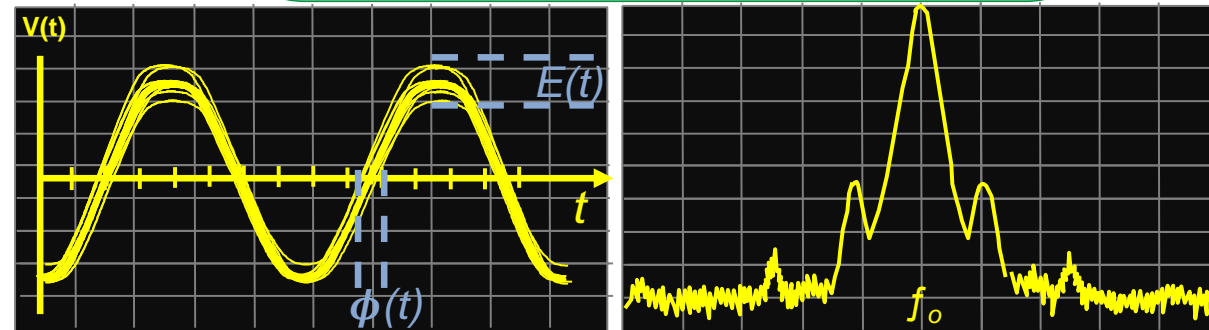
## Real sinusoidal signal

$$V(t) = [A_o + E(t)] \sin[2\pi f_o t + \phi(t)]$$

where

$E(t)$  = random amplitude fluctuations

$\phi(t)$  = random phase fluctuations



Time

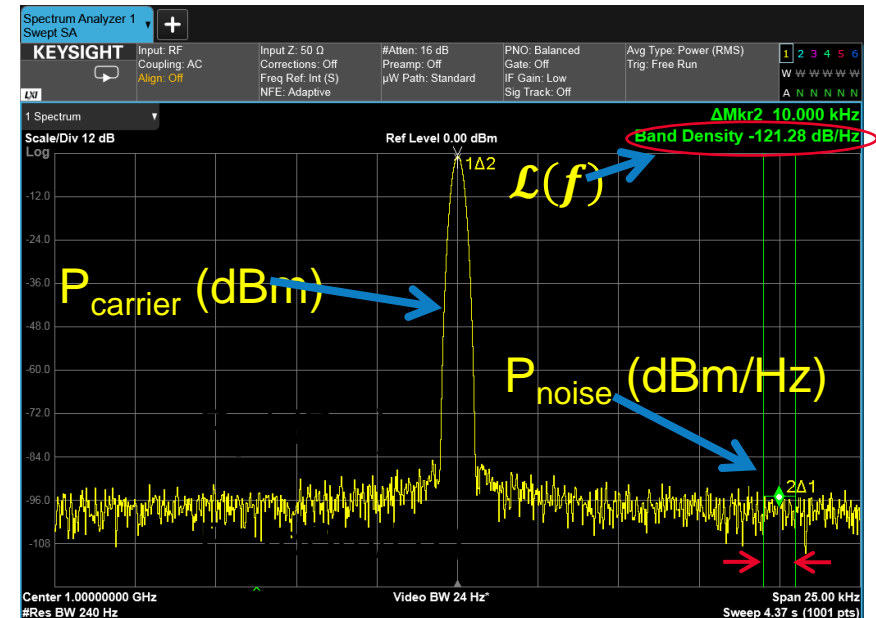
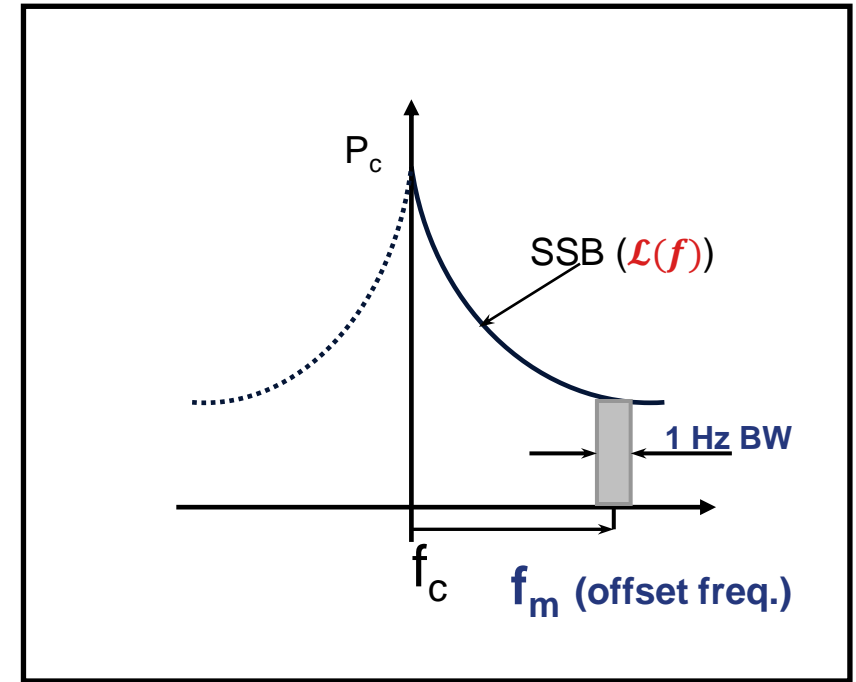
Frequency

# What is Phase Noise?

## How to define phase noise and its measurements

- Phase noise is the phase fluctuation of an oscillator, produced by random noise modulation on its main carrier signal.
- In frequency domain where phase noise is simply noise sidebands or skirt around “ideal” delta function from sinusoidal oscillator, which is symmetrical in magnitude around center frequency, so we can measure a **single noise sideband (SSB)**
- Three elements:
  - **Offset frequency ( $f_m$ ) from carrier frequency ( $f_c$ )**
  - **Power spectral density, in 1 Hz BW**
  - **PSD relative to carrier power in dBc**

$$\mathcal{L}(f) = \frac{P_{SSB}}{P_c} \left( \frac{\text{dBc}}{\text{Hz}} \right)$$



# Spectral Density of Phase Fluctuations

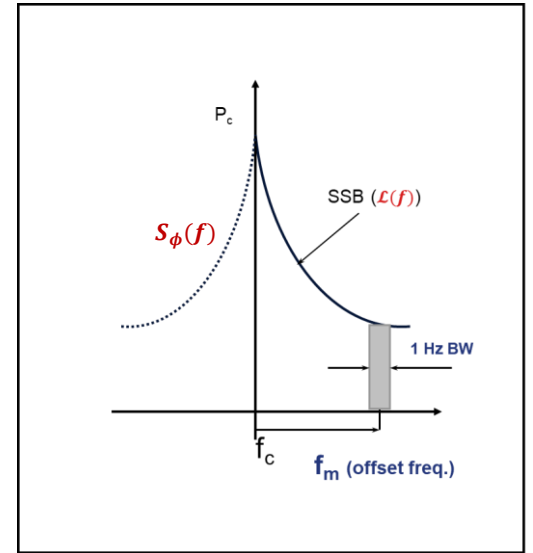
## Mean-square Phase Deviation & Phase Noise

- If we can measure phase deviation ( $\Delta\phi$ ), we can convert mean-square phase deviation measurement (i.e.,  $\phi_{RMS}^2$ ) into a spectral density by dividing by the noise bandwidth and normalizing it to 1 Hz BW:

**Spectral Density of Phase Fluctuations:**  $S_{\phi}(f) = \phi_{RMS}^2 \left(\frac{1}{BW}\right) \left(\frac{rad^2}{Hz}\right)$

- In case of small angle PM (where  $m = \beta = \phi_{pk} \leq 0.2 \text{ rad}$ ), which like we mentioned above in the case of small noise modulates the carrier with small angle deviation in the phase noise situation:

$$\frac{P_{SSB}}{P_c} = \left(\frac{1}{2} \phi_{pk}\right)^2 = \frac{1}{4} (\sqrt{2} \phi_{RMS})^2 = \frac{1}{2} \phi_{RMS}^2 \text{ (rad}^2\text{)}$$



$$\mathcal{L}(f) = \frac{P_{SSB}}{P_c}$$



$$\mathcal{L}(f) \left(\frac{1}{Hz}\right) = \frac{S_{\phi}(f)}{2} = \frac{\phi_{RMS}^2}{2} \left(\frac{1}{BW}\right) \left(\frac{rad^2}{Hz}\right)$$

$$\phi_{RMS} \text{ (rad)} = \sqrt{2 \int_{f_{start}}^{f_{stop}} \mathcal{L}(f) df}$$

# Jitter and Phase Noise

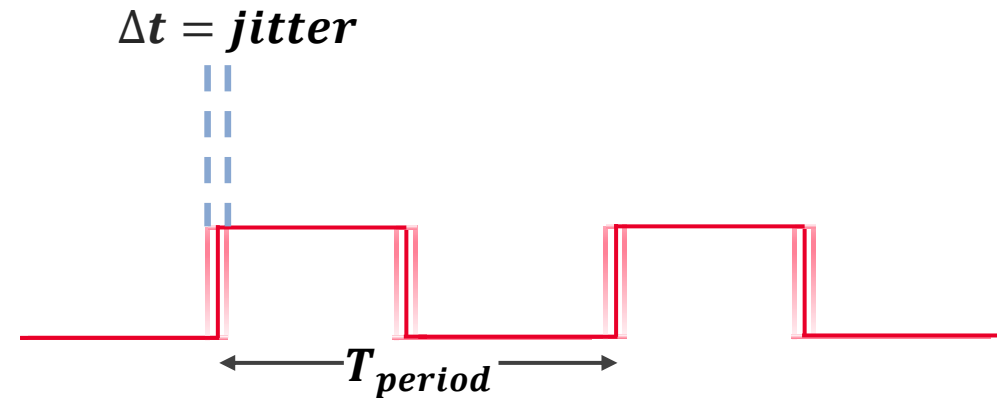
## Jitter definition

- Time-domain RMS phase deviation is called **jitter**
- Those concerned about jitter frequently deal with clock signals, which are usually the measurements of square wave-type signals rather than sinusoids
- The following relates RMS phase deviation to jitter:

$$jitter(\text{seconds}) = \frac{\phi_{RMS}}{2\pi} [T_{period}(\text{seconds})] = \frac{\phi_{RMS}}{2\pi f_c}$$

Percentage of total angular period affected by RMS phase deviation

Clock signal period (time); same as  $1/f_c$



\*  $f_c$  is the clock frequency and when viewed in the frequency domain, is equal to the spacing between the spectral lines

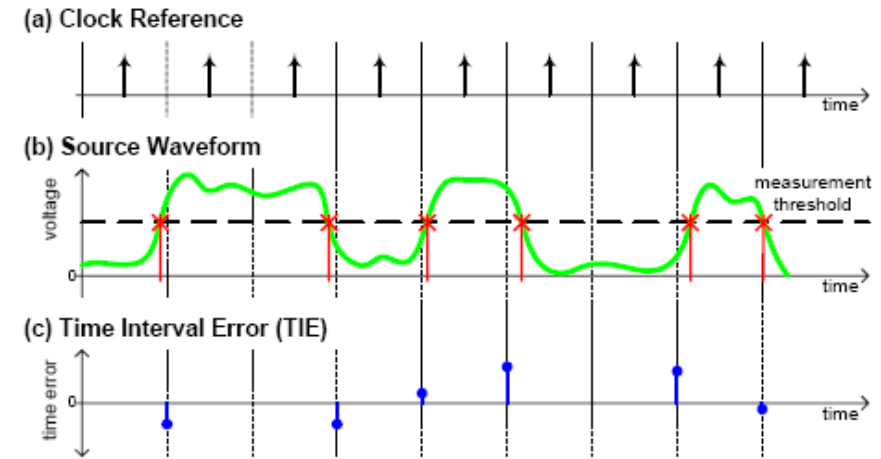
# Jitter and Phase Noise

## Measurement Principles Using a Real-time Oscilloscope

- A **time interval error (TIE)** or **Jitter** measurement on an oscilloscope will produce a time series of the absolute time error of each edge relative to the ideal clock. To convert to phase (radians), the error is simply multiplied by  $2\pi f_c$  where  $f_c$  is the clock carrier frequency:

$$\varphi = 2\pi \cdot TIE(t) \cdot f_c$$

- A **Time Interval Error (TIE) trend** can be transformed to frequency domain with an FFT to give something called a **Jitter Spectrum**. Most modern oscilloscopes have this capability built in or as an option:



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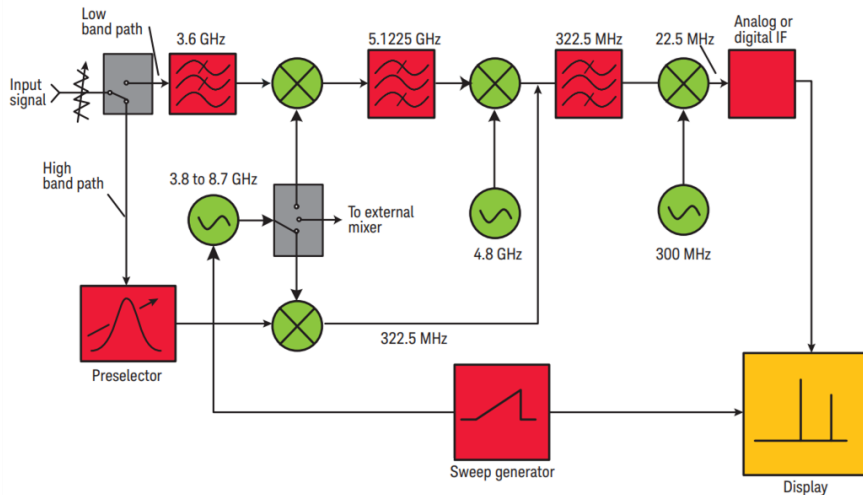
# Measuring Phase Noise in Frequency Domain

## Direct-Spectrum Method

- As mentioned earlier, we use spectrum analyzer to measure the SSB PN:

$$\mathcal{L}(f) = \frac{P_{SSB}}{P_c} \left( \frac{\text{dBc}}{\text{Hz}} \right)$$

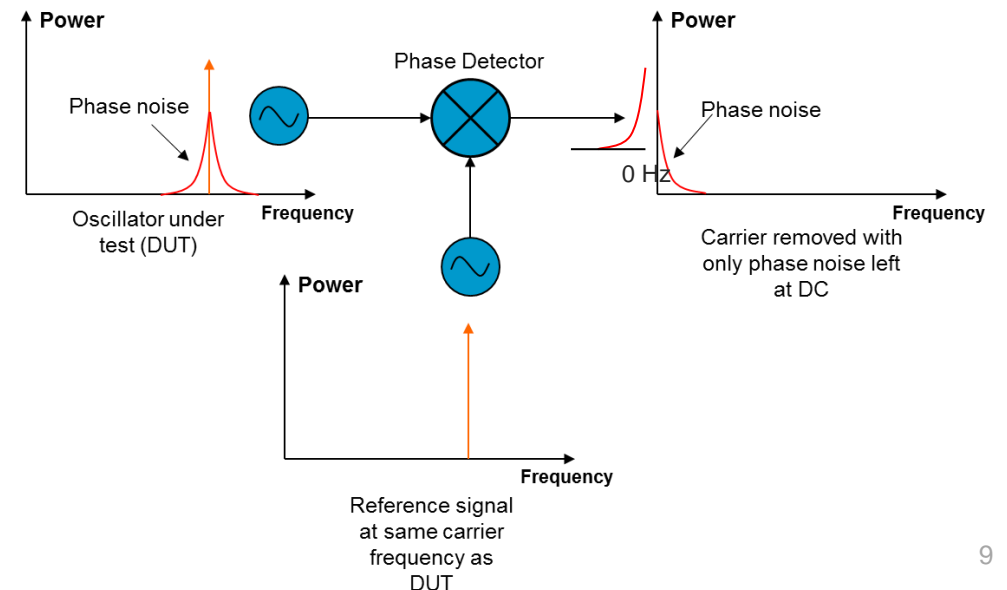
- It is far less sensitive than carrier-removal method because of spectrum analyzer's limitations.
  - Can not separate AM Noise
  - Has LO PN, preamp noise, ADC noise, etc.



## Phase Detector Method

(Carrier-removal method)

- Has increased sensitivity obtained by nulling carrier & then amplifying & measuring phase noise of resulting baseband signal with high-gain, low noise figure amplifiers
- It uses a phase detector to convert phase fluctuation into voltage of noise. We are measuring  $S_{\phi}(f) = \Phi_{RMS}^2 \left( \frac{1}{BW} \right) \left( \frac{\text{rad}^2}{\text{Hz}} \right)$

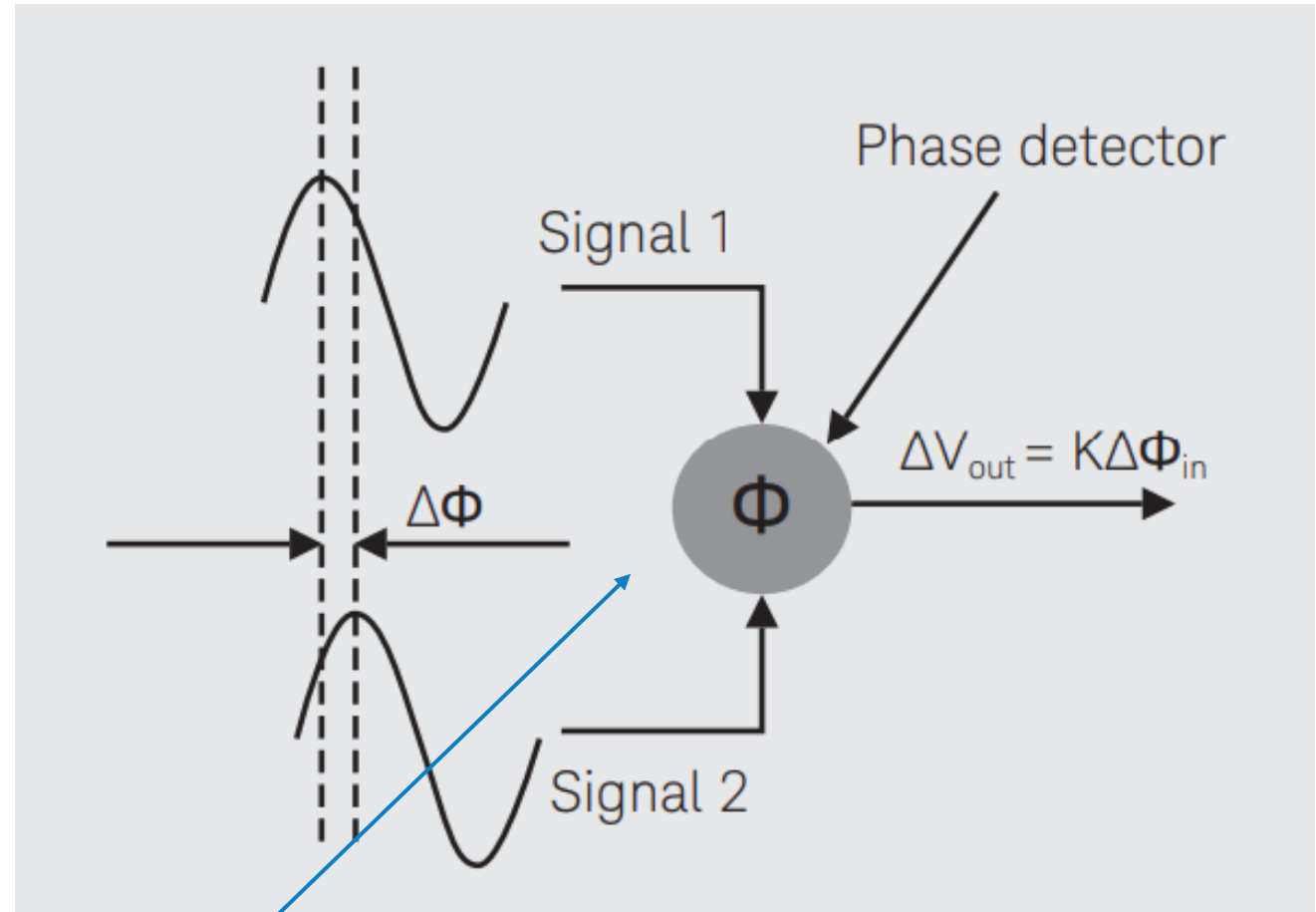


# Phase Detector Techniques

## Carrier-removal method

- Phase detector takes two input signals & compares phase
  - Output of phase detector is DC voltage proportional to delta phase of input signals ( $\Delta\phi$ )
  - Constant of proportionality, K, has units of volts per radian (V/rad) & must be measured
  - When we got the  $\Delta\phi(t)$ , we FFT it to get
- Phase detectors also tend to suppress AM noise

$$S_{\phi}(f) = \phi_{RMS}^2 \left(\frac{1}{BW}\right) \left(\frac{rad^2}{Hz}\right) \text{ and } \mathcal{L}(f) \left(\frac{1}{Hz}\right) = \frac{S_{\phi}(f)}{2}$$

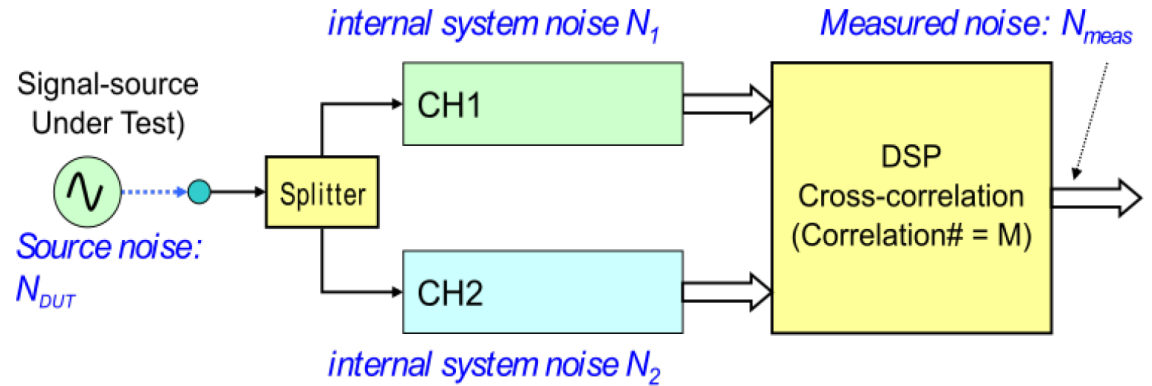


**$\Delta\phi$  to Voltage  
Converter (phase detector)**

# Cross-Correlation Technique

- We can use two phase detectors to further improve phase noise floor (i.e., sensitivity)
- Two channels are uncorrelated so remove noise from references & system components through computational process.
- DUT signal is common to both channels so is perfectly correlated in both channels & kept as measurement result
- The two-channel cross-correlation technique achieves superior measurement sensitivity without requiring exceptional performance of the hardware components.
- However, the measurement speed suffers when increasing the number of computational correlations.

Two-Channel Cross-Correlation Technique:



$$N_{meas} = N_{DUT} + (N_1 + N_2) / \sqrt{M} \quad \text{Assuming } N_1 \text{ and } N_2 \text{ are uncorrelated.}$$

M (number of correlation)	10	100	1,000	10,000
Noise reduction on $(N_1+N_2)$	-5dB	-10dB	-15dB	-20dB

DUT noises through each channel are coherent and are not affected by the cross-correlation, whereas the internal noises generated by each channel are incoherent and are diminished by the cross-correlation operation at the rate of  $M^{1/2}$  (M being the number of correlations). This can be expressed as:

$$N_{meas} = N_{DUT} + (N_1 + N_2) / \sqrt{M}$$

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# Keysight Phase Noise Solution Portfolio



## N5511A Phase Noise Test System (PNTS)

*Fast and flexible noise measurement at the limit of physics ( $kT = -177$  dBm/Hz)*

- Absolute / residual / AM / baseband / pulse / transient
- 50 kHz to 3.0 / 26.5 / 40 GHz
- 0.01 Hz to 160 MHz offset frequency
- Cross correlation
- E5500 replacement (100% code compatible)
- External mixer support for mmWave measurements



Legacy SSA (E5052B) +  
Downconverter (E5053A)



Feb'24 Release



New

## E505xA Signal Source Analyzer (SSA-X)

*Everything you need for signal source analysis*

- Absolute / residual / 2-port VNA (S-parameters) / AM / baseband / pulse / transient / SA / Clock jitter
- 1 MHz to 8 / 26.5 / 44 / 54 GHz
- >54 GHz with external mixers (sub-THz)
- 1 mHz to > 1 GHz offset frequency
- Cross correlation

Building on 35 years of phase noise experience, Keysight solutions provide excellent performance and are economically tailored to fit your needs.



## X-series Signal Analyzers

### N9068 Phase Noise Measurement Application

- Absolute / residual / AM / baseband / pulse / transient
- Frequency range, offset range and performance depends on X-series model

## PNA/PNA-X Network Analyzers

### S9303xxB Phase Noise Application

- Absolute / residual / AM / baseband / pulse / transient
- 10 MHz to 70 / 125 GHz
- 0.1 Hz to 10 MHz offset frequency

## Infinium / MXR / UXR Oscilloscopes

### D90x0JITA Jitter, Vertical and Phase Noise Analysis

- Absolute / residual / AM / baseband / pulse / transient
- Frequency range, offset range and performance depends on oscilloscope model
- Cross correlation
- Best for measurements on digital signals, such as clocks

Performance ↑

Ease of Use →

[www.keysight.com/find/phasenoise](http://www.keysight.com/find/phasenoise)

# Thank you