

# Performance of ICARUS T600 electronics at LNGS and its upgrade for operation on SBN at FNAL.

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# ICARUS-WA104 Collaboration

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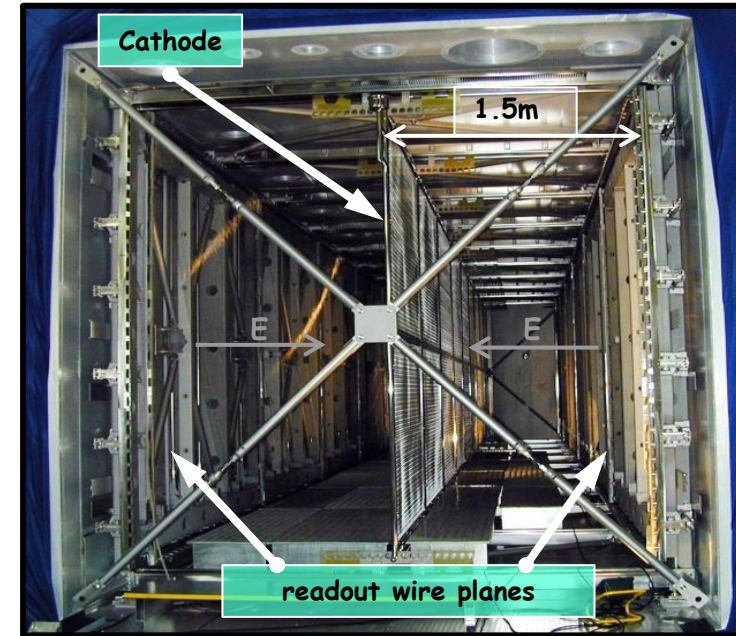
Eight American Institutions: ANL, BNL, FNAL, LANL, SLAC, Univ. of Colorado, Univ. of Pittsburgh, Univ. of Texas

- ICARUS T600 TPC
- ICARUS T600 electronics and DAQ
- Upgrading for SBN operation at FNAL
- Cosmic rays test results
- Conclusions

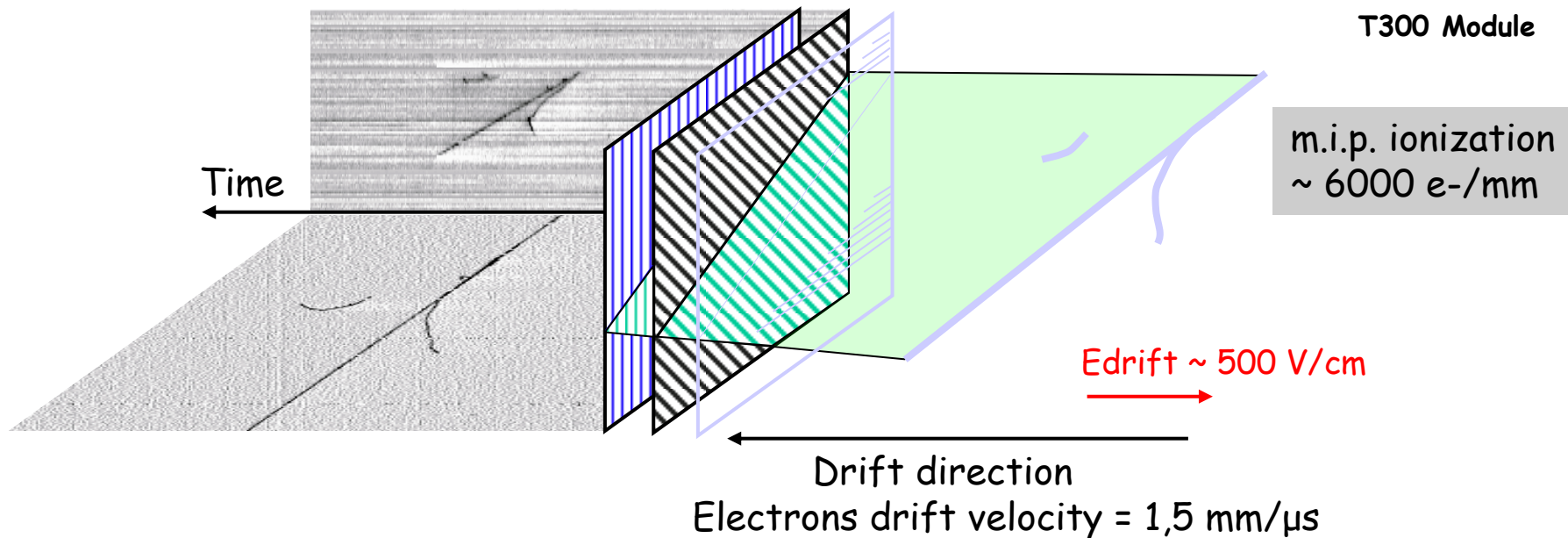
# ICARUS T600 TPC

ICARUS T600 is the *unique* example of very large mass liquid argon (LAr) Time Projection Chamber (TPC). It provides 3D imaging of any ionizing event (like an electronic bubble chamber). A major feature is the continuous sensitivity, self triggering capability, and calorimetric measurement.

- ❖ 2 identical T300 modules adjacent (3.6m x 3.9m x 19.6m each)
- ❖ 2 chambers per module, 1.5 m drift length each
- ❖ 3 readout wire planes per chamber wires at  $0, \pm 60^\circ$  (ind1, ind2, coll view)
- ❖ 53248 wires, 3 mm pitch and plane spacing



T300 Module



# ICARUS front-end electronics racks



96 racks on the top of detector:

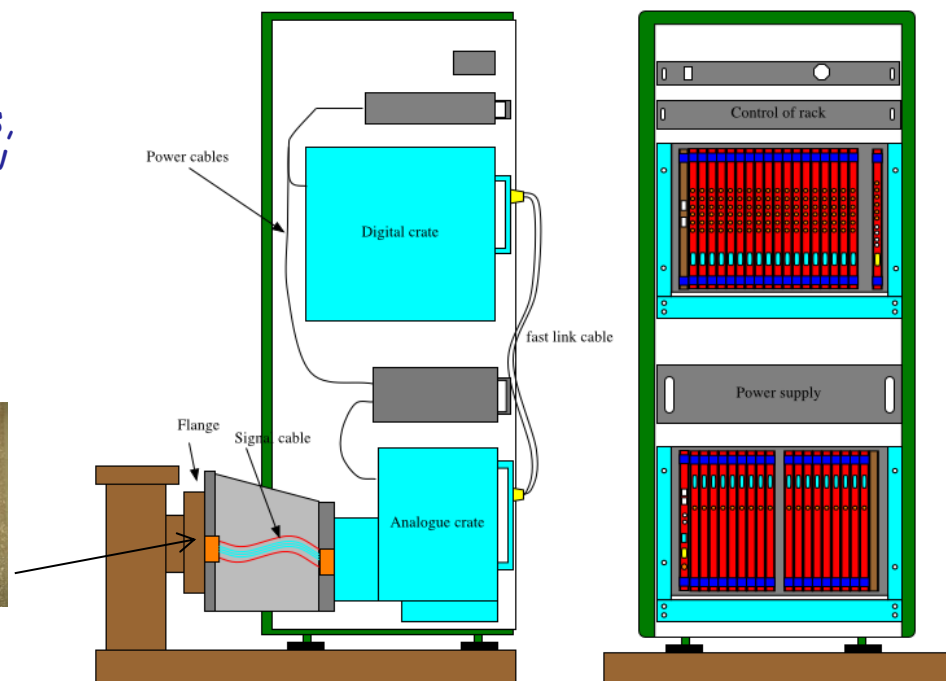
- ❖ 16 racks for the horizontal wires of Induction-1 plane (528 ch per rack).
- ❖ 72 racks for the wires at  $\pm 60^\circ$  of Induction-2 and collection plane (576 ch per rack, except of four racks, first and last of each row, with 448 ch each).
- ❖ 8 racks for the wires of decreasing length at  $\pm 60^\circ$  from the corners of the wire frame (544 ch per rack).

Each single rack, in front of each feedthrough flange, houses 576 readout channels (32 ch modularity) and contains:

- ❖ One VME-like analogue crate with 18 analog boards, 18 decoupling boards in the backplane, and one slow control Mod.
- ❖ One digital VME crate with 18 digital boards, one CPU, and one clock/trigger distributor Mod.
- ❖ Power supplies and their control and monitor.



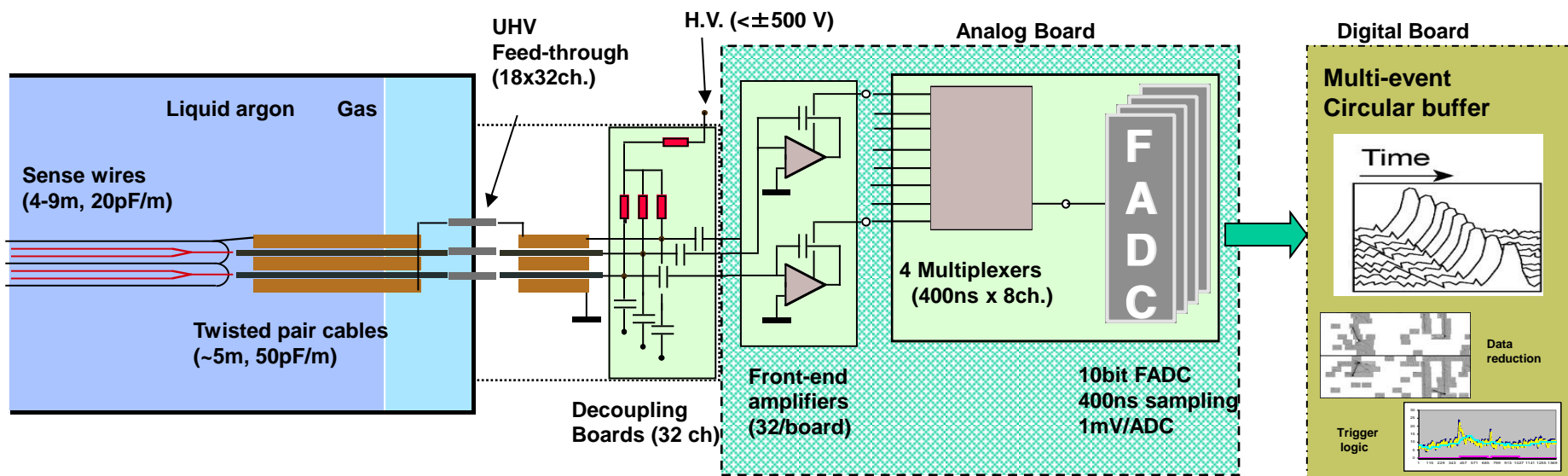
UHV Feed-through  
(18 x 32ch.)





# ICARUS front-end electronics

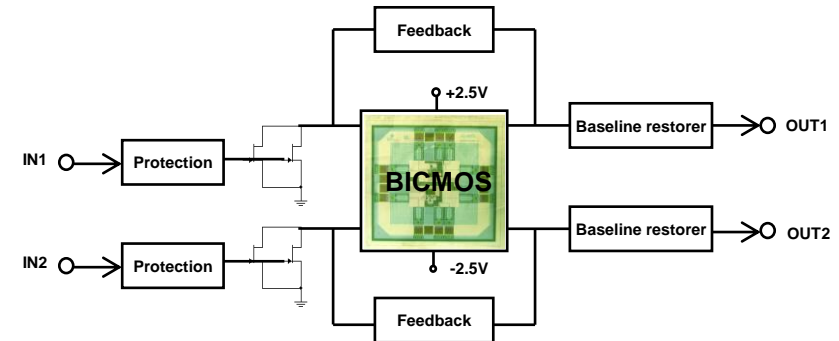
- ❖ The ICARUS T600 read-out electronics was designed to provide continuous digitization and waveform recording of the signals from each wire of the TPC.
- ❖ **Decoupling Board:** it receives 32 analogue signals from the chamber and passes them to the analogue board via decoupling capacitors; it also provides wire biasing voltage and distribution of the test signals.
- ❖ **Analog Board:** it hosts 32 front-end low noise charge sensitive pre-amplifiers, performs data multiplexing and data conversion ADC (10 bit). The sampling period for each channel is 400 ns.
- ❖ **Digital Board:** it provides multi event buffer memory for 32 channels, data compression, and trigger logic.



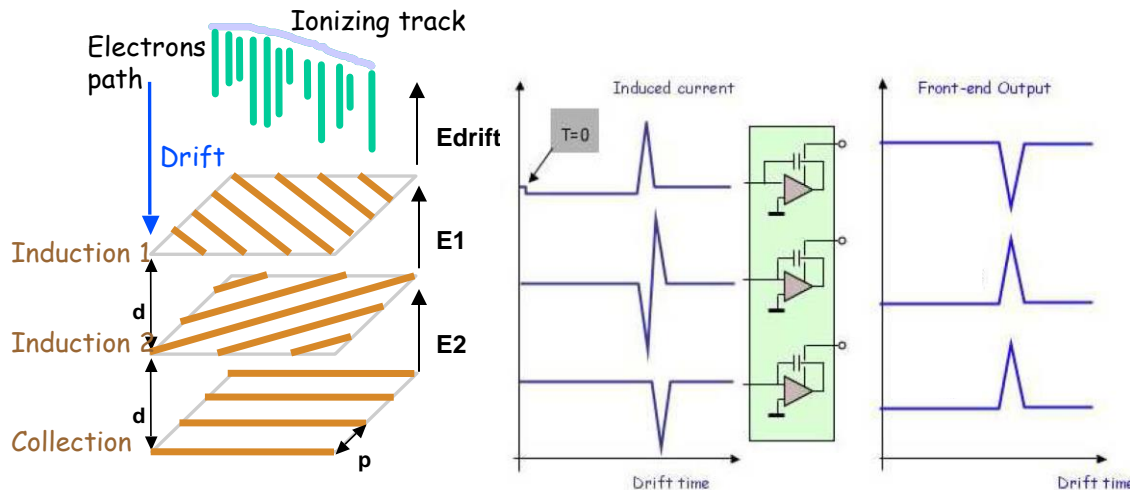
- ❖ Signal to noise ratio (S/N) better than 10 and a  $\sim 0.6$  mm single point resolution were obtained during the LNGS run, resulting in precise spatial reconstruction of events, allowing for measuring muon momentum by multiple scattering (MS) with  $\Delta p/p \sim 16\%$  in the 0.4-4 GeV/c range.

# ICARUS preamplifier and signals

- ❖ Input stage: two jFet are connected in parallel.
- ❖ Custom IC in BiCMOS technology
  - Two channels unfolded Radeka integrator
  - External feedback network
- ❖ Baseline restorer circuit
- ❖ Two versions
  - "quasi-current" mode:  $R_f C_f \approx 3\mu s$  (collection + first induction);
  - "quasi-charge" mode:  $R_f C_f \approx 100\mu s$  (mid induction);



Sensitivity  $\approx 6 \text{ mV/fC}$   
Dynamic range  $> 200 \text{ fC}$   
Linearity  $< 0.5\%$  @ full scale  
Gain uniformity  $< 3\%$   
E.N.C.  $\approx (350 + 2.5 \times C_D) \text{ el}$   
 $\approx 1200 \text{ el. @ } 350\text{pF}$   
Power consumption  $\approx 40 \text{ mW/channel}$



## Shaping time at output:

- ❖ Current mode (Induction 1, Collection):  $\sim 3\mu s$ . Signal proportional to induced current. Integral of collection signal area proportional to charge.
- ❖ Charge mode (Induction 2):  $\sim 3\mu s$ . Signal amplitude proportional to charge.

# Lossless data compression

## Online lossless data compression

- ❖ In data collected with T600 LAr-TPC, the difference between one sample and the previous one is within  $\pm 7$  ADC counts in more than about 98% of the cases.
- ❖ This allows for storing the differences instead of the full 10 bit data, using fewer bits.
- ❖ Assuming to handle data in 2-Byte format, the choice is to pack four 4 bit difference ( $\pm 7$  ADC counts) obtaining a  $\sim 4$  compression factor.
- ❖ When the difference is larger than  $|7|$ , the full difference is stored in 2-Byte with a 4 bit flag (1000).
- ❖ The compression efficiency is affected by the large energy deposition from e.m showers or high dE/dx tracks.
- ❖ During LNGS run the real measured compression factor was 3.92.

| Compression scheme                  |    |    |    |                                       |    |   |   |                                       |   |   |   |                                       |   |   |   |
|-------------------------------------|----|----|----|---------------------------------------|----|---|---|---------------------------------------|---|---|---|---------------------------------------|---|---|---|
| 15                                  | 14 | 13 | 12 | 11                                    | 10 | 9 | 8 | 7                                     | 6 | 5 | 4 | 3                                     | 2 | 1 | 0 |
| 4-bit<br>Difference of<br>channel N |    |    |    | 4-bit<br>Difference of<br>channel N+1 |    |   |   | 4-bit<br>Difference of<br>channel N+2 |   |   |   | 4-bit<br>Difference of<br>channel N+3 |   |   |   |

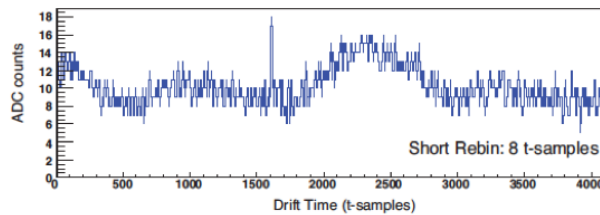
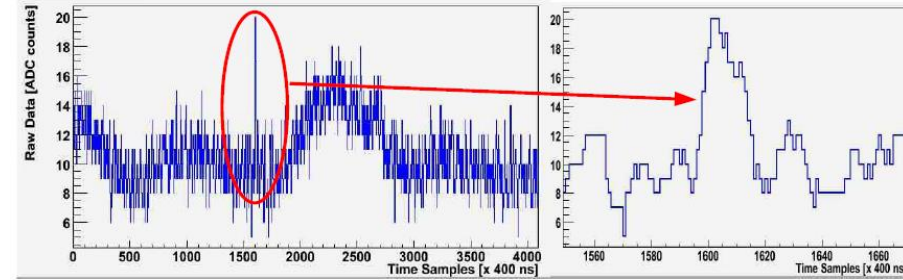
| Difference |    |    |    |    |    |   |   |                        |   |   |   |   |   |   |   |
|------------|----|----|----|----|----|---|---|------------------------|---|---|---|---|---|---|---|
| 15         | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7                      | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| 1 0 0 0    |    |    |    |    |    |   |   | 10-bit full difference |   |   |   |   |   |   |   |



# Hit finding algorithm (double rebinning sliding windows)

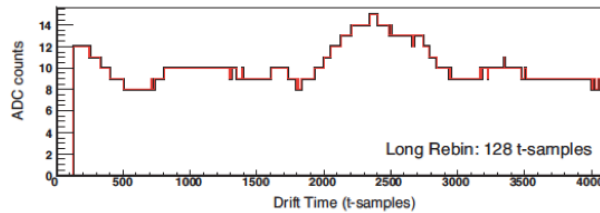
Real time algorithm for hit finding implemented on FPGA:

- ❖ A typical m.i.p. signal: ~15 ADC counts, 30-40 t-samples, could be affected by:
- ❖ Low frequency noise (fluctuation of the baseline): 10-15 ADC counts, 1500 t-samples;
- ❖ High frequency noise: 2-3 ADC counts, 5 t-samples.



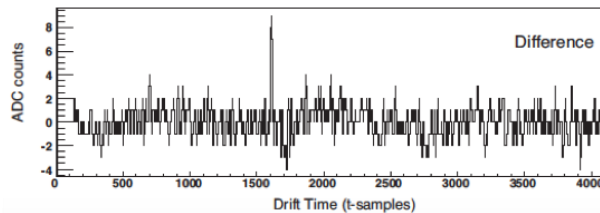
The high frequency component can be smoothed by averaging the waveform over a short time window.

$$Q_8(t) = \frac{1}{8} \sum_{i=0}^8 Q(t-i)$$



The low frequency component can be smoothed by averaging the waveform over a long time window.

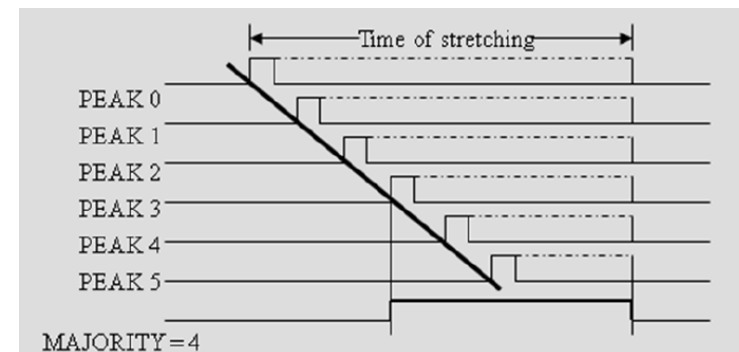
$$Q_{128}(t) = \frac{1}{128} \sum_{i=0}^{128} Q(t-i)$$



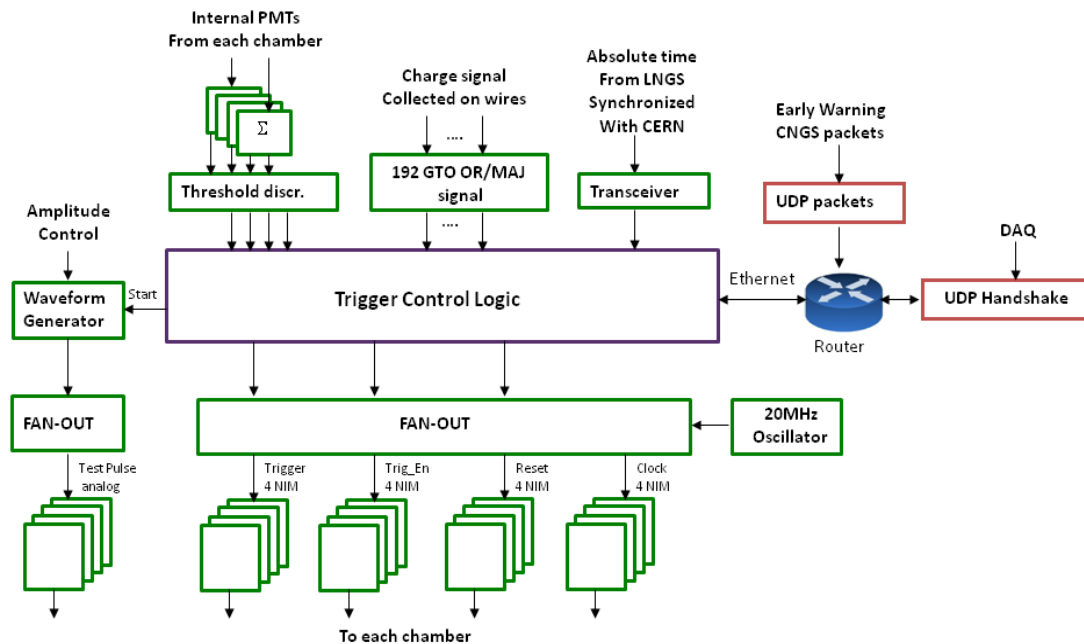
A hit signal is generated when the difference  $S(t)$  goes over threshold.

$$S(t) = Q_8(t) - Q_{128}(t)$$

- ❖ Majority over 16 consecutive wires (~5cm).
- ❖ Peak stretching (25-125  $\mu$ s) to guarantee high efficiency majority selection for inclined tracks.
- ❖ A local trigger is generated by the logical OR of the two majority signal on the same board.



# ICARUS global trigger scheme



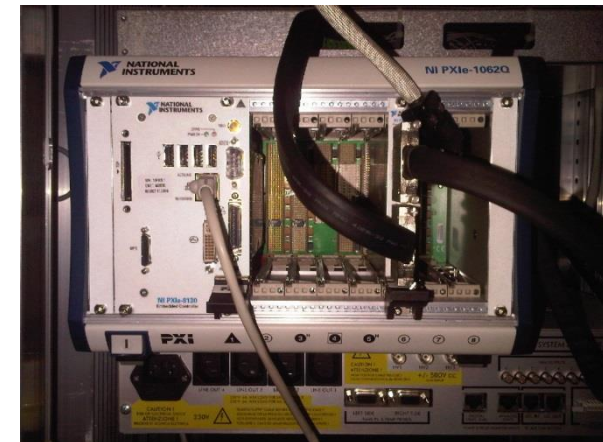
- ❖ A continuous communication with DAQ, in handshake mode, prevents the generation of new triggers in the case of the detector busy, while a multi-veto configuration minimizes dead-time.

## Trigger sources

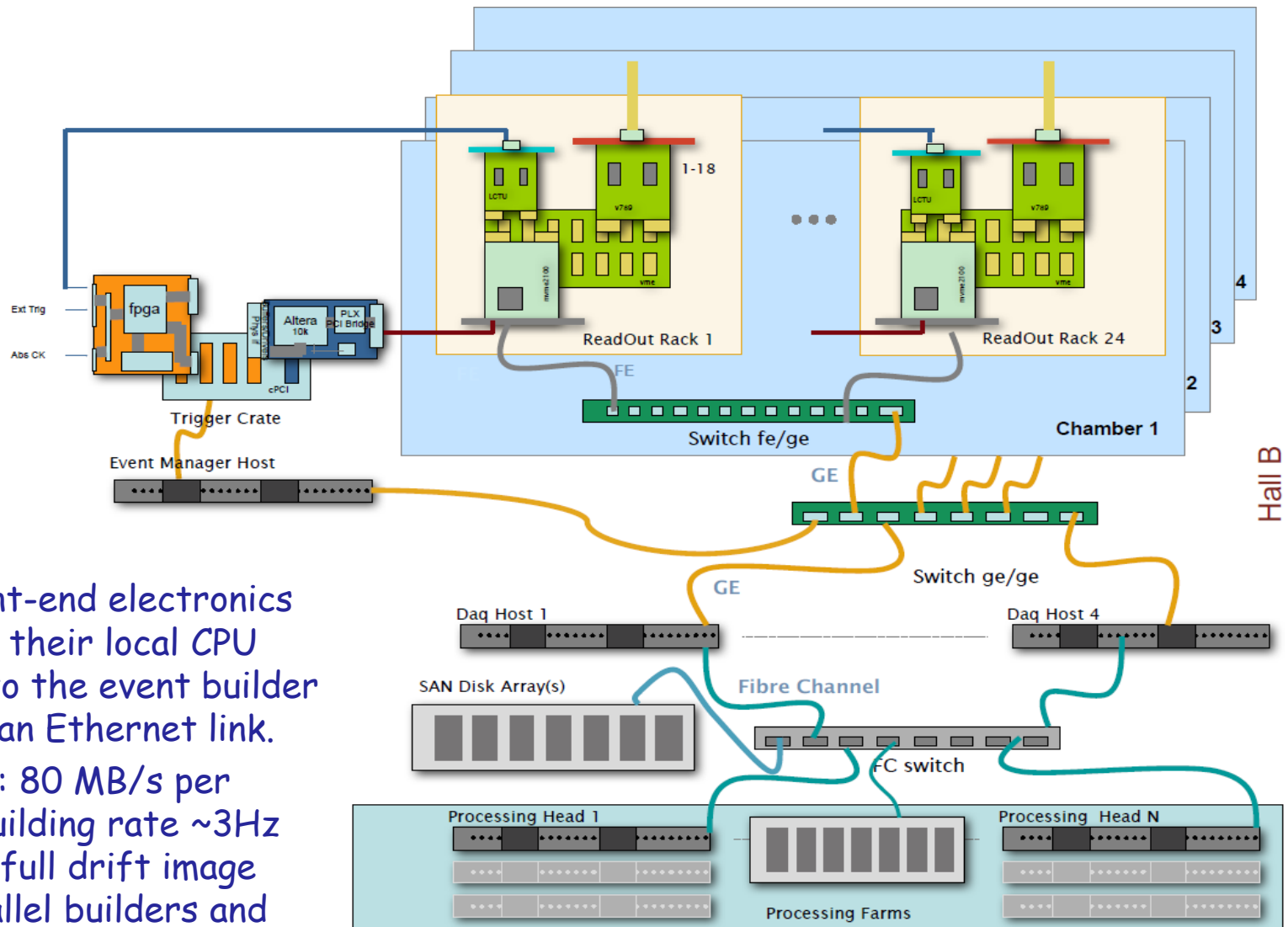
- ❖ CNGS beam gate (obtained from Early warning through internet and GPS);
- ❖ Light signals collected by 74 PMTs;
- ❖ Charge signals (TPC wires).

## Trigger implementation

- ❖ Commercial NI PXI crate;
- ❖ Real time controller (PXIe-8130): trigger - DAQ communication;
- ❖ FPGA board (PXI-3813R, PXI-7833): signal handling, time critical processing.



# ICARUS T600 DAQ event builder

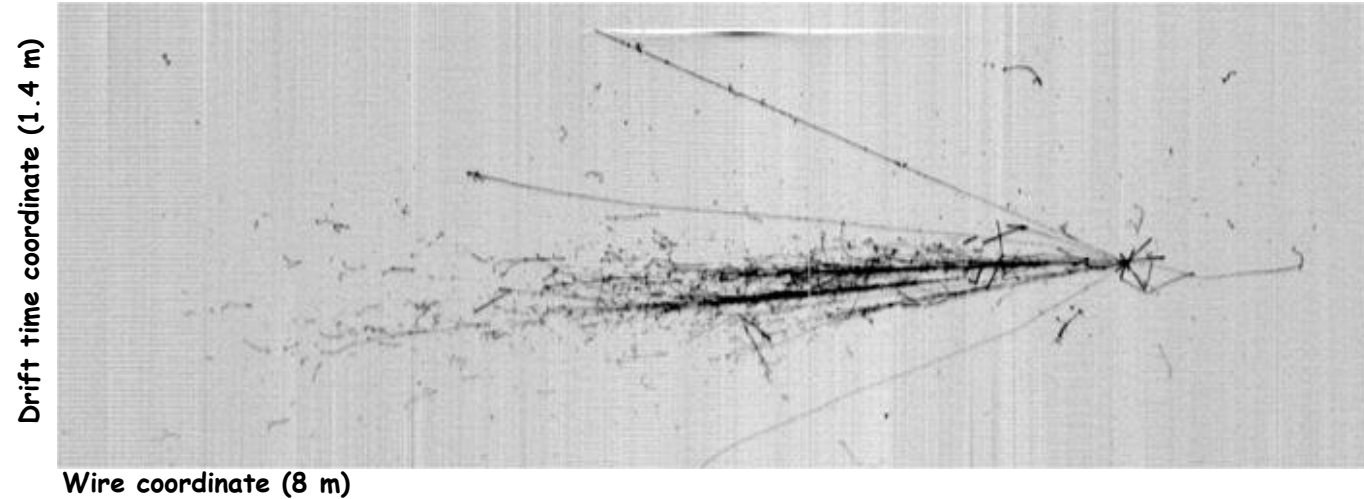


- ❖ The 96 front-end electronics crates have their local CPU connected to the event builder PC through an Ethernet link.
- ❖ Throughput: 80 MB/s per chamber, building rate  $\sim 3\text{Hz}$  for a 1.6ms full drift image using 4 parallel builders and data compression factor of 4.

# CNGS neutrino interaction in ICARUS T600 (May 2010)

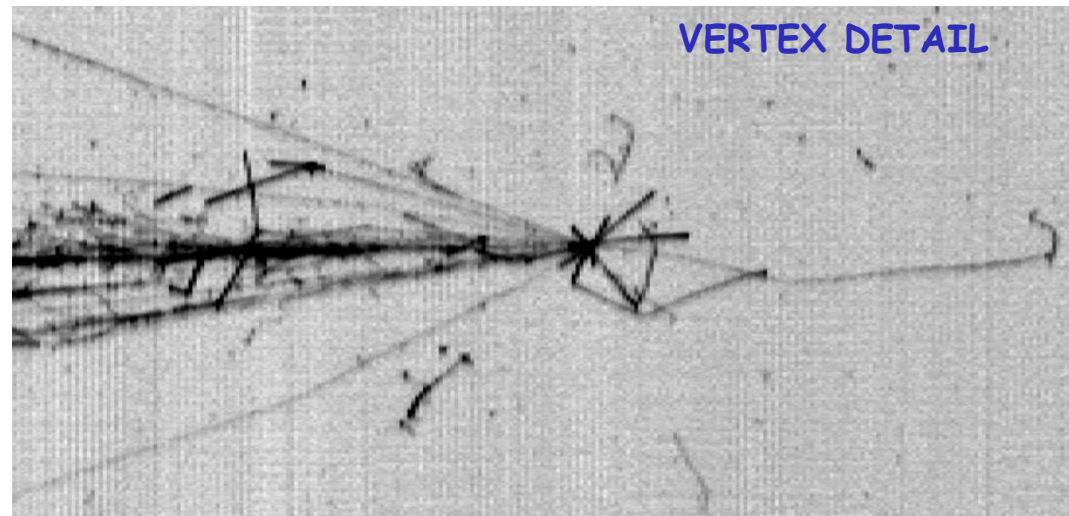
Collection view

CNGS neutrino beam direction  
←



**The first neutrino event!**  
**0.6 mm spatial resolution**  
**in 8m image size.**

Extremely high quality image  
thanks to the very high S/N  
of electronics (not forgetting  
purity, mechanical precision  
and stability...).



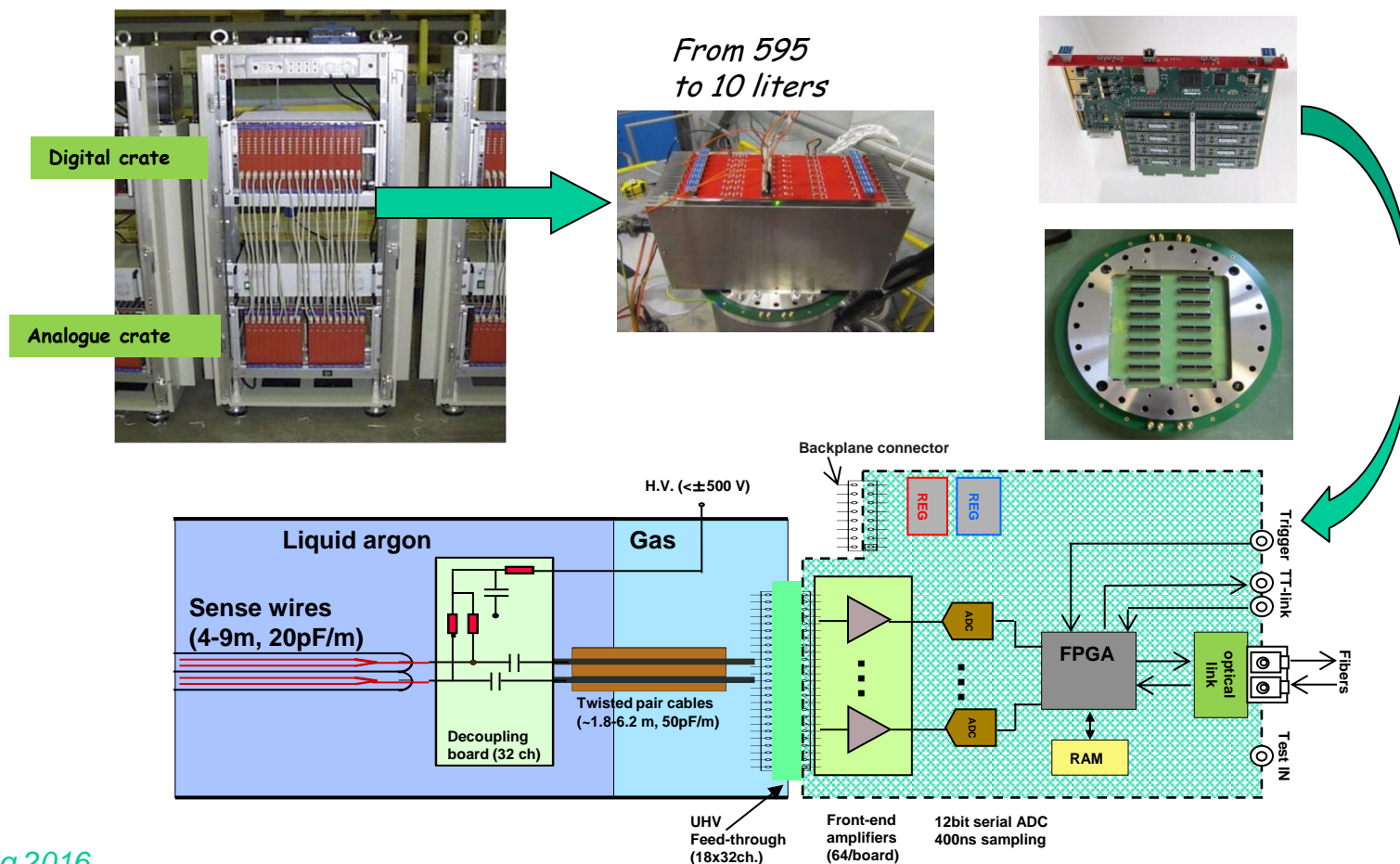
# The need for an upgraded electronics

- ❖ Even though old electronics can be considered state of the art for the time it was conceived (1998) and well suited for ICARUS-T600, this electronics has some limitations:
  - poor treatment of induction signals, in case of showers or high  $dE/dx$  tracks, due to signal undershoot;
  - limited data throughput due to the choice of VME standard (8-10 MB/s), perfectly legitimate at that time.
  
- ❖ Improvements concern:
  - short shaping peak time of the analogue signals to avoid undershoot;
  - adoption of serial ADCs and smaller packages of components to make the system much more compact;
  - adoption of a modern serial bus architecture with optical links (Gbit/s) to increase bandwidth;
  - use of the flange as electronics backplane to simplify layout and cabling.



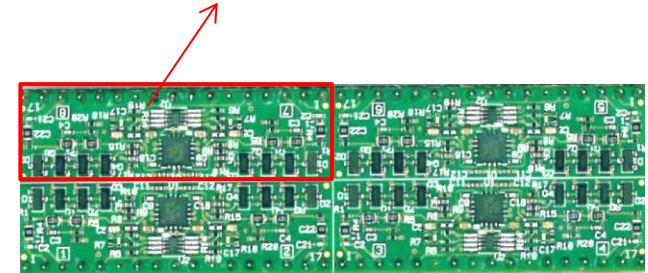
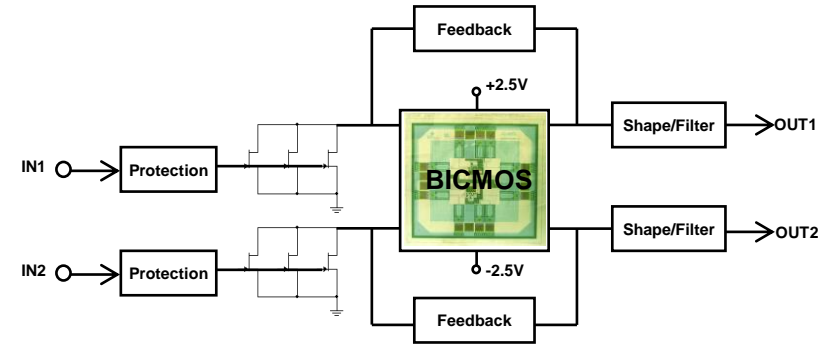
# New simplified/compact design

- ❖ A new, compact design, has been conceived to host both analogue and digital electronics on a single board directly connected to the proprietary flanges.
- ❖ One mini-crate, mounted on the flange, can host 9 boards for 576 channels, 64 channels each.
- ❖ The backplane of the crate distributes the power supply and local control signals.
- ❖ A single boards hosts 64 front-end low noise charge sensitive pre-amplifiers, 64 serial 12 bit ADC (2.5 MHz), FPGA, memory, optical link interface...

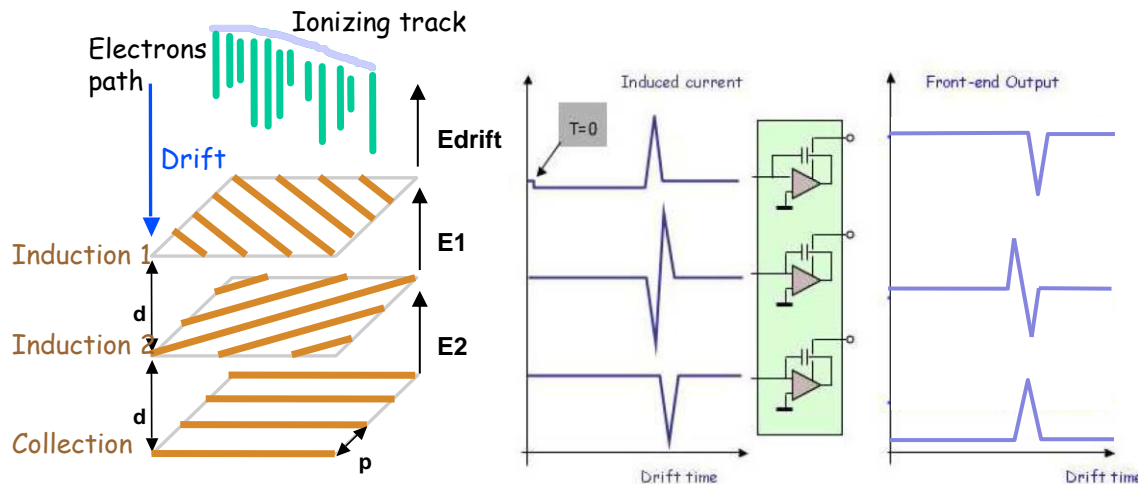


# Improved preamplifier

- ❖ Three jFet are connected in parallel to increase  $g_m$  (50-60 mS).
- ❖ Same pre-amp response (shaping peak time and gain) both for collection and induction signals.
- ❖ Adoption of a smaller package for the custom BiCMOS dual channel amplifier.
- ❖ The gain of the front-end amplifier and filter is  $3\text{ V}/300\text{ fC}$ . The 12bit ADC input range is  $3.3\text{ V}$  with a least count equivalent to  $\sim 500$  electrons. This value matches with the amplifier noise of  $\sim 1000$  electrons with an "equivalent" detector capacitance of  $\sim 270\text{ pF}$  (wires plus cables).



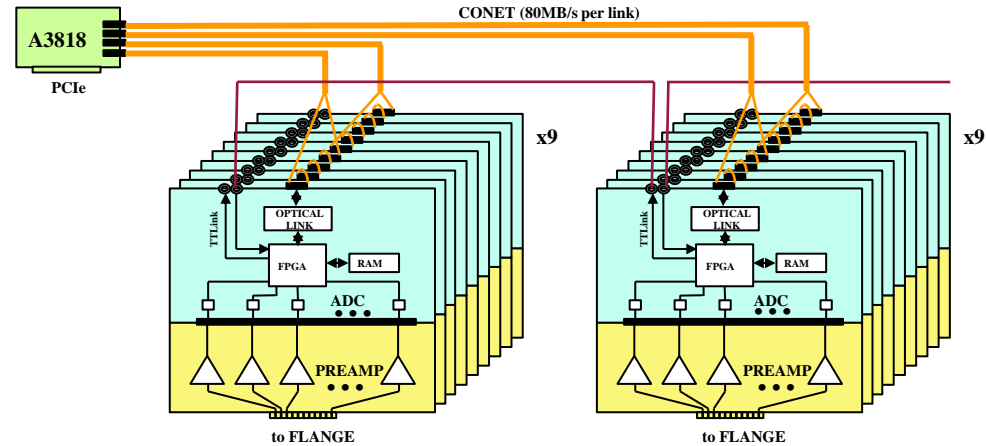
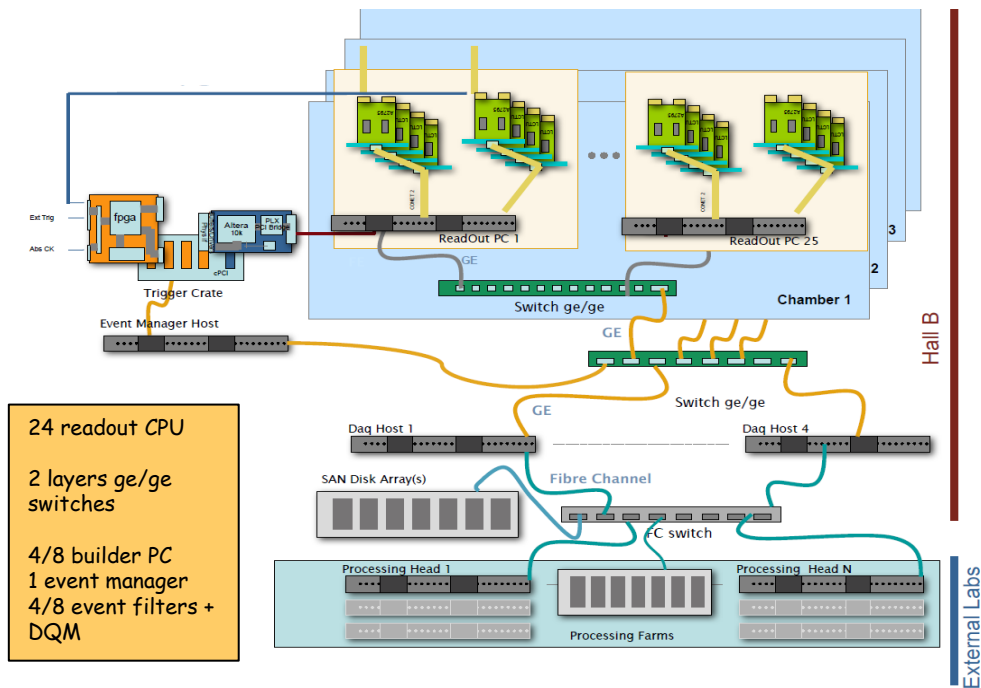
8-channel Preamplifier module



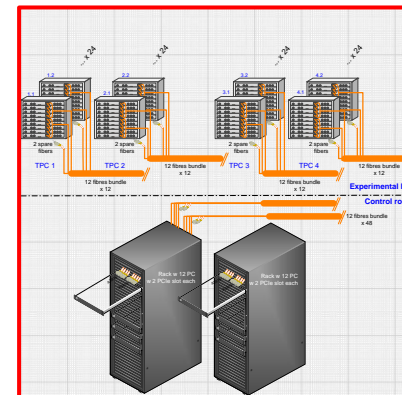
- ❖ All views (including Ind2) are read out with pole-zero cancellation circuit, shaping peak time  $\tau = 1.5\mu\text{s}$ .
- ❖ Short shaping peak time preserves bipolar signals allowing for numerical integration of the digitized output.

# Upgrading DAQ

- ❖ The system provisionally uses the CONET transfer protocol.
- ❖ Each mini-crate (flange) will require two CONET loops.
- ❖ Each A3818 can handle 4 loops (2 flanges).
- ❖ On each PC can host 2 A3818, a total of 24 readout PC will be needed for the whole detector



- ❖ The readout DAQ could keep the existing DAQ architecture, simply replacing the VME CPU in each readout unit with a PC equipped with a CONET interface. Expected building rate  $\sim 15\text{Hz}$  without data compression.



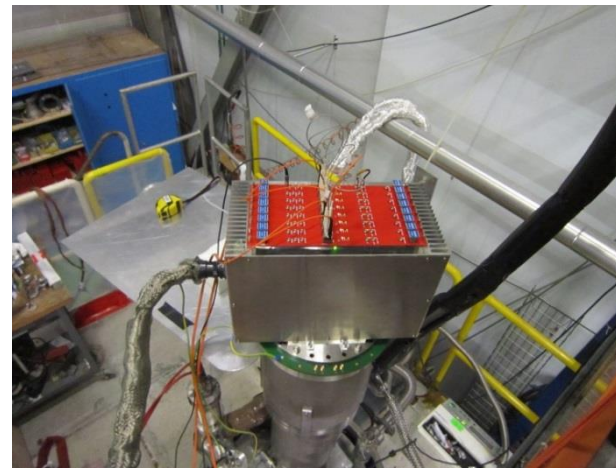
- ❖ The whole DAQ can be hosted in a 54U rack
- ❖ 4 X 24 fiber bundles (+ spares) from control room to mini crate (~50/100m)



# Test facility in CERN (50 liter LAr TPC)

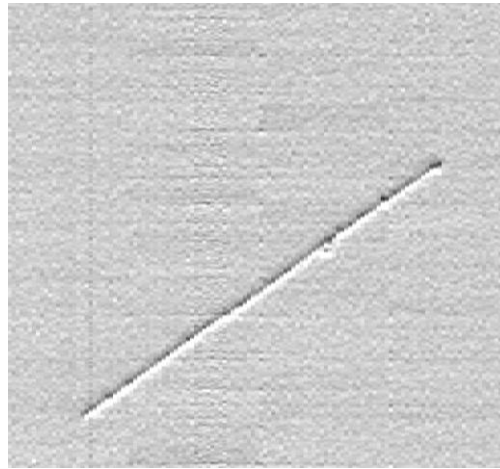
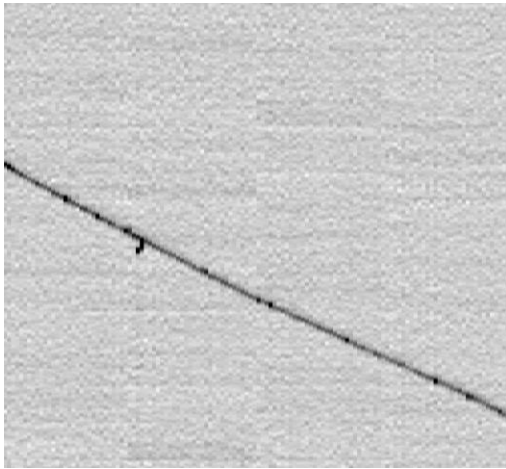


- ❖ 2 wire planes (ind, coll), 128 wires each;
- ❖ 2.54 mm wire pitch and 4 mm plane spacing
- ❖ 46.8 cm drift length, at 500V/cm.
- ❖ Cable length as in ICARUS collection plane, about 2.5m



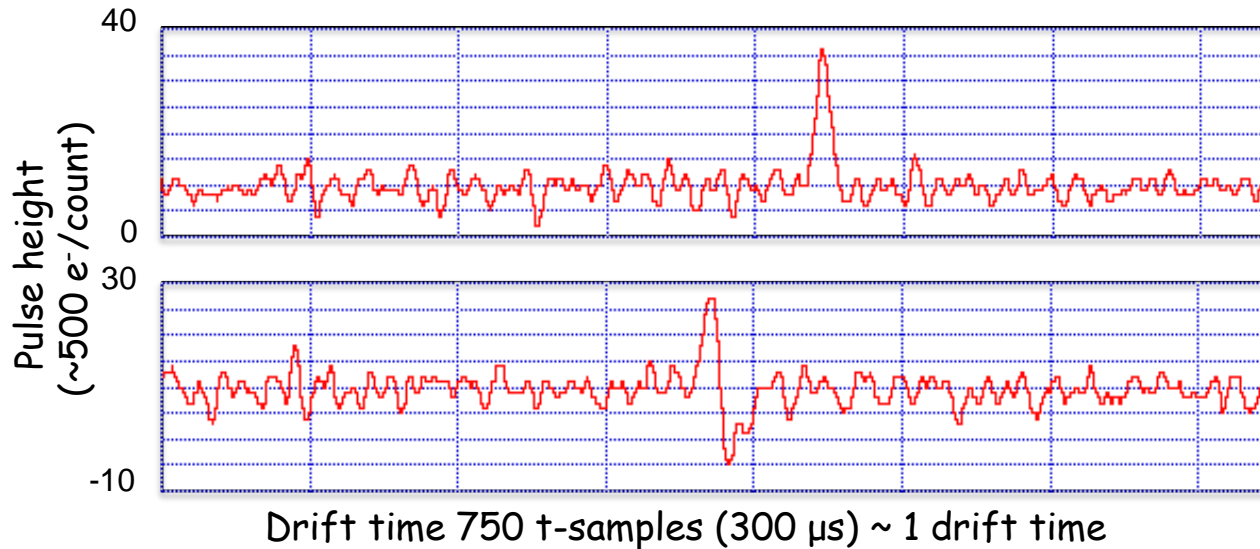
# Cosmic rays test @ 50 liter CERN LAr TPC

300  $\mu\text{s}$  (465 mm)  
Drift velocity 1.55 mm/ $\mu\text{s}$



128 collection wires (325 mm)    128 induction wires (325 mm)

- ❖ A single mip track event:
  - Same  $\sim 2$ ADC counts ( $\sim 1000$   $e^-$ ) noise for both Collection & Induction;
  - unipolar collection signal:  $\sim 25$  ADC counts;
  - Symmetric bipolar induction signal with slightly reduced amplitude as expected.
  - No filter applied to any data.

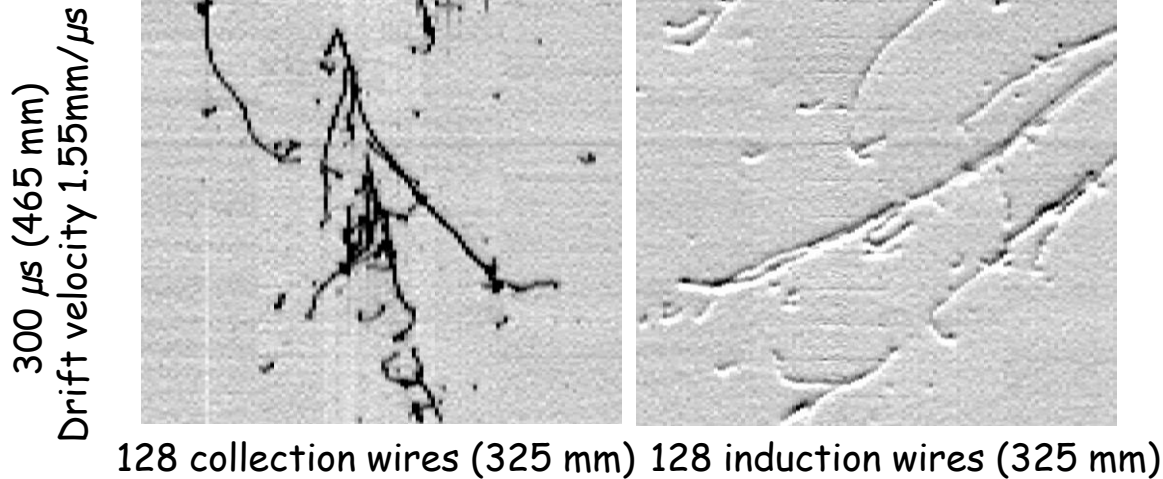


collection signal (on a single wire)

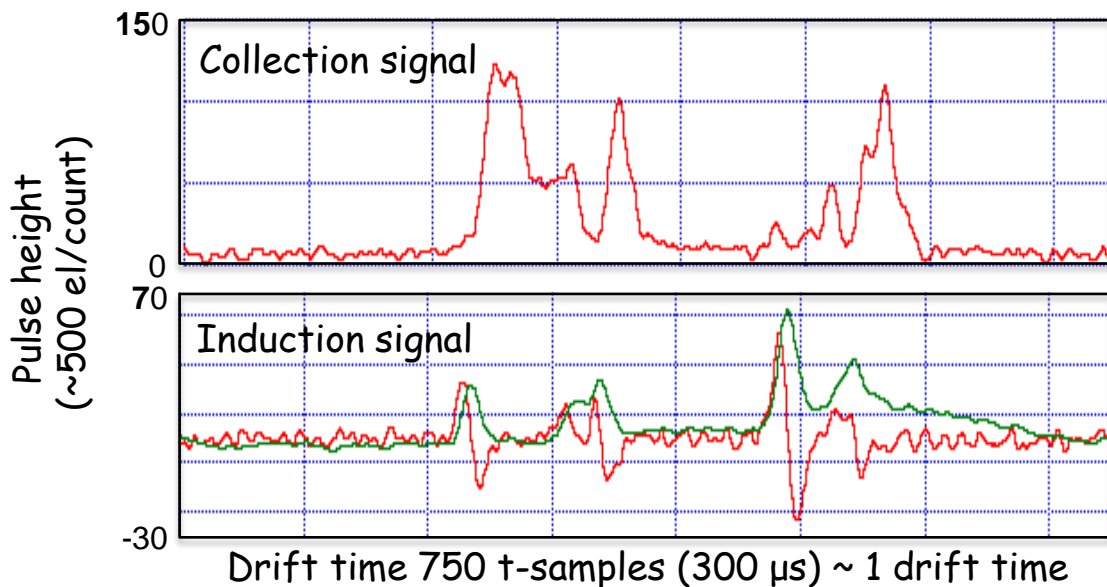
induction signal (on a single wire)



# Cosmic rays test @ 50 liter CERN LAr TPC

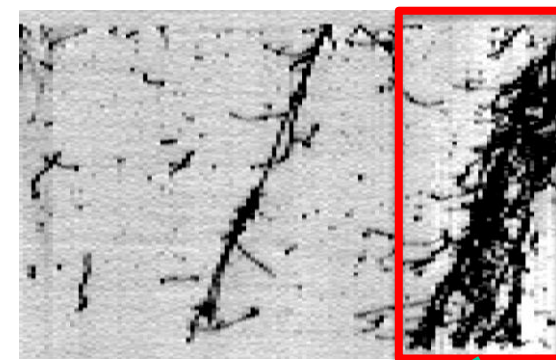
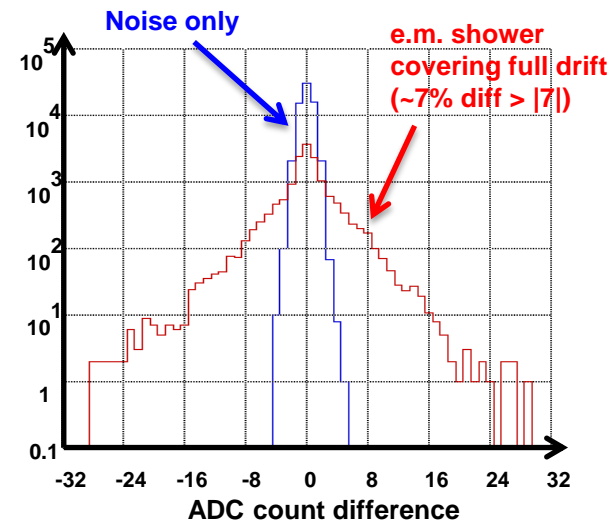


- ❖ A shower event developing along the drift direction in Collection;
- ❖ The optimized preamp architecture results in:
  - no signal undershoot even after large signals;
  - a very stable baseline;
  - unprecedented image sharpness and better hit position separation due to the faster shaping peak time.
- ❖ On induction plane, energy information easily recoverable with dedicated algorithms (e.g. running sum, green curve).
- ❖ No filter applied to any data.



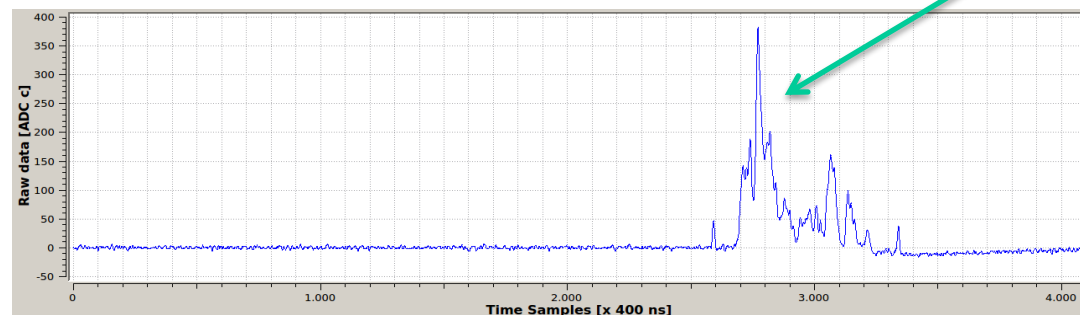
# Lossless data compression

- ❖ From the test with the 50 Liter LAr-TPC, due to the low rms noise ( $\sim 1000$  electrons = 2 ADC counts), the difference between consecutive samples is within  $\pm 7$  ADC counts in 100% of the cases in absence of tracks.
- ❖ In case of an event, as shown inside the red rectangle (very high density shower), the difference distribution is as in the red curve above the event image: only 7% of the cases the difference exceeds  $\pm 7$  ADC counts.
- ❖ We can then use the same compression scheme used for the old system based on 10 bit ADC, according to the model in the figure, expecting a compression factor very similar to the previous one.



| Compression scheme            |    |    |    |                                 |    |   |   |                                 |   |   |   |                                 |   |   |   |
|-------------------------------|----|----|----|---------------------------------|----|---|---|---------------------------------|---|---|---|---------------------------------|---|---|---|
| 15                            | 14 | 13 | 12 | 11                              | 10 | 9 | 8 | 7                               | 6 | 5 | 4 | 3                               | 2 | 1 | 0 |
| 4-bit Difference of channel N |    |    |    | 4-bit Difference of channel N+1 |    |   |   | 4-bit Difference of channel N+2 |   |   |   | 4-bit Difference of channel N+3 |   |   |   |

| Difference |    |    |    |                        |    |   |   |   |   |   |   |   |   |   |   |
|------------|----|----|----|------------------------|----|---|---|---|---|---|---|---|---|---|---|
| 15         | 14 | 13 | 12 | 11                     | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| 1          | 0  | 0  | 0  | 12-bit full difference |    |   |   |   |   |   |   |   |   |   |   |



# Conclusions

- ❖ ICARUS T600 is a milestone being the largest LAr TPC ever operated underground. It produced some thousands high quality neutrino events thanks to its mechanical precision and stability, LAr purity, and electronics quality.
- ❖ T600 TPC overhauling is in progress at CERN in view of its transfer to FNAL on SBN beam.
- ❖ The improved electronic read out will allow for using also induction view for dE/dx measurement.
- ❖ Better S/N ratio (>12).
- ❖ The full synchronization of the sampling time in whole system will improve the muon momentum measurement by MS (reduction of  $\Delta p/p$  from 16% to 13% ).
- ❖ The data throughput will increase to ~15Hz.
- ❖ Embedding the signal boards onto the feedthrough flange makes the system extremely compact, without external cables between flange and electronics.
- ❖ Signal cables do not carry high voltage resulting in microphonic noise reduction.
- ❖ Fully external electronics makes the system easy to maintain and suitable for any future evolution.





Thank you !